

μA776

MULTI-PURPOSE PROGRAMMABLE OPERATIONAL AMPLIFIER

FAIRCHILD LINEAR INTEGRATED CIRCUIT

DESCRIPTION — The μA776 Programmable Operational Amplifier is constructed using the Fairchild Planar* epitaxial process. High input impedance, low supply currents, and low input noise over a wide range of operating supply voltages coupled with programmable electrical characteristics result in an extremely versatile amplifier for use in high accuracy, low power consumption analog applications. Input noise voltage and current, power consumption, and input current can be optimized by a single resistor or current source that sets the chip quiescent current for nano-watt power consumption or for characteristics similar to the μA741. Internal frequency compensation, absence of latch-up, high slew rate and short circuit current protection assure ease of use in long time integrators, active filters, and sample and hold circuits.

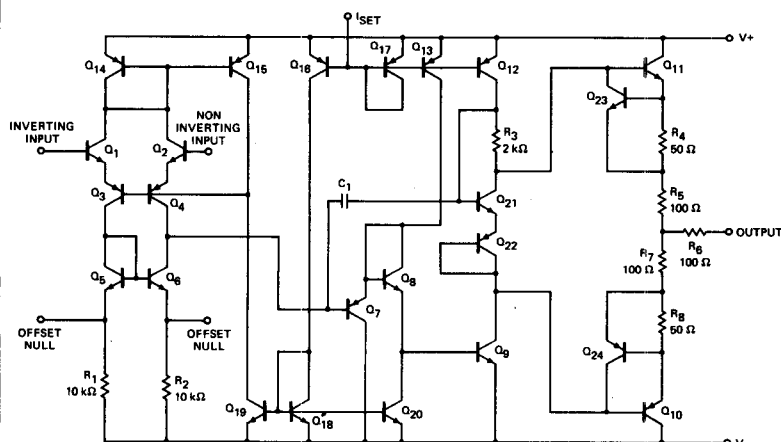
- MICROPOWER CONSUMPTION
- $\pm 1.2V$ to $\pm 18V$ OPERATION
- NO FREQUENCY COMPENSATION REQUIRED
- LOW INPUT BIAS CURRENTS
- WIDE PROGRAMMING RANGE

- HIGH SLEW RATE
- LOW NOISE
- SHORT CIRCUIT PROTECTION
- OFFSET NULL CAPABILITY
- NO LATCH-UP

ABSOLUTE MAXIMUM RATINGS

Supply Voltage	$\pm 18 V$
Internal Power Dissipation (Note 1)	
Metal Can	500 mW
DIP	670 mW
Mini DIP	310 mW
Differential Input Voltage	$\pm 30 V$
Input Voltage (Note 2)	$\pm 15 V$
Voltage Between Offset Null and V-	$\pm 0.5 V$
I _{SET} (Maximum Current at I _{SET})	500 μA
V _{SET} (Maximum Voltage to Ground at I _{SET})	$(V+ - 2.0 V) < V_{SET} < V+$
Storage Temperature Range	
Metal Can, DIP	$-65^{\circ}C$ to $+150^{\circ}C$
Mini DIP	$-55^{\circ}C$ to $+125^{\circ}C$
Operating Temperature Range	
Military (μA776)	$-55^{\circ}C$ to $+125^{\circ}C$
Commercial (μA776C)	$0^{\circ}C$ to $+70^{\circ}C$
Lead Temperature (Soldering, 60 seconds)	
Metal Can, DIP	$300^{\circ}C$
Mini DIP	$260^{\circ}C$
Output Short Circuit Duration (Note 3)	Indefinite

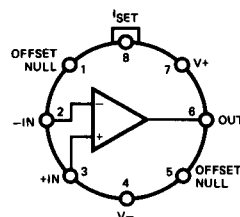
EQUIVALENT CIRCUIT



CONNECTION DIAGRAMS

8-LEAD METAL CAN (TOP VIEW)

PACKAGE OUTLINE 5S PACKAGE CODE H

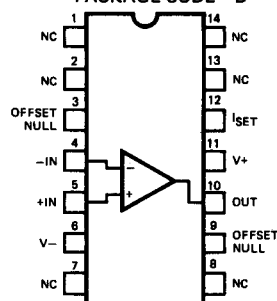


ORDER INFORMATION

TYPE	PART NO.
μA776	μA776HM
μA776C	μA776HC

14-LEAD DIP (TOP VIEW)

PACKAGE OUTLINE 6A PACKAGE CODE D

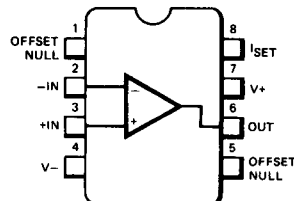


ORDER INFORMATION

TYPE	PART NO.
μA776	μA776DM
μA776C	μA776DC

8-LEAD MINI DIP (TOP VIEW)

PACKAGE OUTLINE 9T PACKAGE CODE T



ORDER INFORMATION

TYPE	PART NO.
μA776C	μA776TC

*Planar is a patented Fairchild process.

FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A776$

± 15 V OPERATION FOR $\mu A776$

ELECTRICAL CHARACTERISTICS ($T_A = 25^\circ\text{C}$, unless otherwise specified)

PARAMETERS	CONDITIONS	$I_{SET} = 1.5\mu\text{A}$			$I_{SET} = 15\mu\text{A}$			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	$R_S \leq 10\text{k}\Omega$		2.0	5.0		2.0	5.0	mV
Input Offset Current	$R_S \leq 10\text{k}\Omega$		0.7	3.0		2.0	15	nA
Input Bias Current			2.0	7.5		15	50	nA
Input Resistance			50			5.0		M Ω
Input Capacitance			2.0			2.0		pF
Offset Voltage Adjustment Range			9.0			18		mV
Large Signal Voltage Gain	$R_L \geq 75\text{k}\Omega$, $V_{OUT} = \pm 10\text{V}$	200k	400k					V/V
	$R_L \geq 5\text{k}\Omega$, $V_{OUT} = \pm 10\text{V}$				100k	400k		V/V
Output Resistance			5.0k			1.0k		Ω
Output Short-Circuit Current			3.0			12		mA
Supply Current			20	25		160	180	μA
Power Consumption				0.75			5.4	mW
Transient Response (unity gain)	$V_{IN} = 20\text{mV}$, $R_L \geq 5\text{k}\Omega$, $C_L = 100\text{pF}$		1.6			0.35		μs
			0			10		%
Slew Rate	$R_L \geq 5\text{k}\Omega$		0.1			0.8		V/ μs
Output Voltage Swing	$R_L \geq 75\text{k}\Omega$	± 12	± 14					V
	$R_L \geq 5\text{k}\Omega$				± 10	± 13		V

The following specifications apply $-55^\circ\text{C} < T_A < +125^\circ\text{C}$

Input Offset Voltage	$R_S \leq 10\text{k}\Omega$			6.0			6.0	mV
Input Offset Current	$T_A = +125^\circ\text{C}$			5.0			15	nA
	$T_A = -55^\circ\text{C}$			10			40	nA
Input Bias Current	$T_A = +125^\circ\text{C}$			7.5			50	nA
	$T_A = -55^\circ\text{C}$			20			120	nA
Input Voltage Range		± 10			± 10			V
Common Mode Rejection Ratio	$R_S \leq 10\text{k}\Omega$	70	90		70	90		dB
Supply Voltage Rejection Ratio	$R_S \leq 10\text{k}\Omega$		25	150		25	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L \geq 75\text{k}\Omega$, $V_{OUT} = \pm 10\text{V}$	100k			75k			V/V
Output Voltage Swing	$R_L \geq 75\text{k}\Omega$	± 10			± 10			V
Supply Current				30			200	μA
Power Consumption				0.9			6.0	mW

FAIRCHILD LINEAR INTEGRATED CIRCUITS • $\mu A776$

± 3 V OPERATION FOR $\mu A776$

ELECTRICAL CHARACTERISTICS ($T_A = 25^\circ\text{C}$, unless otherwise specified)

PARAMETERS	CONDITIONS	$I_{SET} = 1.5\mu\text{A}$			$I_{SET} = 15\mu\text{A}$			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	$R_S \leq 10k\Omega$		2.0	5.0		2.0	5.0	mV
Input Offset Current			0.7	3.0		2.0	15	nA
Input Bias Current			2.0	7.5		15	50	nA
Input Resistance			50			5.0		M Ω
Input Capacitance			2.0			2.0		pF
Offset Voltage Adjustment Range			9.0			18		mV
Large Signal Voltage Gain	$R_L > 75k\Omega$, $V_{OUT} = \pm 1V$	50k	200k					V/V
	$R_L > 5k\Omega$, $V_{OUT} = \pm 1V$				50k	200k		V/V
Output Resistance			5k			1k		Ω
Output Short-Circuit Current			3.0			5.0		mA
Supply Current			13	20		130	160	μA
Power Consumption			78	120		780	960	μW
Transient Response (unity gain)	Rise Time	$V_{IN} = 20\text{mV}$, $R_L \geq 5k\Omega$, $C_L < 100\text{pF}$				0.6		μs
	Overshoot					5		%
Slew Rate	$R_L > 5k\Omega$		0.03			0.35		V/ μs

The following specifications apply for $-55^\circ\text{C} < T_A < +125^\circ\text{C}$

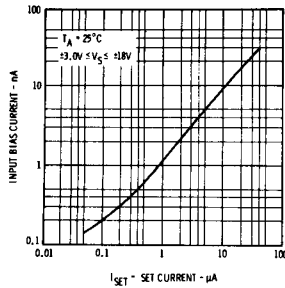
Input Offset Voltage	$R_S \leq 10k\Omega$			6.0			6.0	mV
Input Offset Current	$T_A = +125^\circ\text{C}$			5.0			15	nA
	$T_A = -55^\circ\text{C}$			10			40	nA
Input Bias Current	$T_A = +125^\circ\text{C}$			7.5			50	nA
	$T_A = -55^\circ\text{C}$			20			120	nA
Input Voltage Range		± 1.0			± 1.0			V
Common Mode Rejection Ratio	$R_S \leq 10k\Omega$	70	86		70	86		dB
Supply Voltage Rejection Ratio	$R_S \leq 10k\Omega$		25	150		25	150	$\mu\text{V/V}$
Large Signal Voltage Gain	$R_L > 75k\Omega$, $V_{OUT} = \pm 1V$	25k						V/V
	$R_L > 5k\Omega$, $V_{OUT} = \pm 1V$				25k			V/V
Output Voltage Swing	$R_L > 75k\Omega$	± 2.0	± 2.4					V
	$R_L > 5k\Omega$				± 1.9	± 2.1		V
Supply Current				25			180	μA
Power Consumption				150			1080	μW

NOTES:

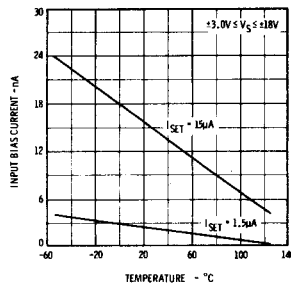
- Rating applies to ambient temperatures up to 70°C . Above 70°C ambient derate linearly at $6.3 \text{ mW}/^\circ\text{C}$ for Metal Can, $8.3 \text{ mW}/^\circ\text{C}$ for the DIP, and $5.6 \text{ mW}/^\circ\text{C}$ for the Mini DIP.
- For supply voltages less than $\pm 15 \text{ V}$, the absolute maximum input voltage is equal to the supply voltage.
- Short Circuit may be to ground or either supply. Rating applies to $+125^\circ\text{C}$ case temperature or $+75^\circ\text{C}$ ambient temperature for $I_{SET} \leq 30 \mu\text{A}$.

TYPICAL PERFORMANCE CURVES FOR $\mu A776$ AND $\mu A776C$

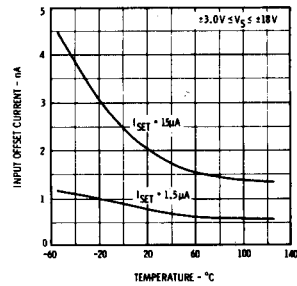
INPUT BIAS CURRENT
AS A FUNCTION OF
SET CURRENT



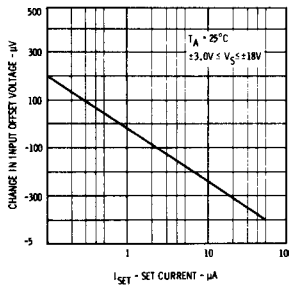
INPUT BIAS CURRENT
AS A FUNCTION OF
AMBIENT TEMPERATURE



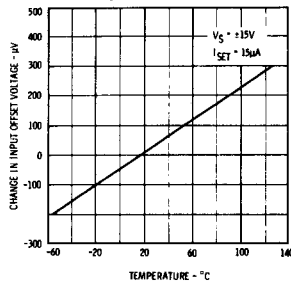
INPUT OFFSET CURRENT
AS A FUNCTION OF
AMBIENT TEMPERATURE



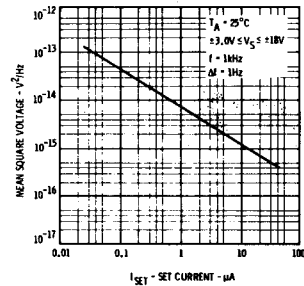
CHANGE IN INPUT OFFSET
VOLTAGE AS A FUNCTION
OF SET CURRENT



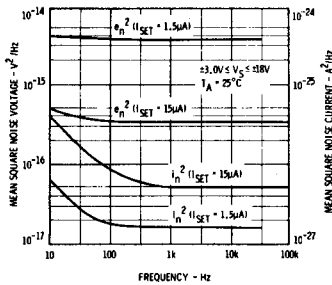
CHANGE IN INPUT OFFSET
VOLTAGE AS A FUNCTION
OF AMBIENT TEMPERATURE
(UNNULLED)



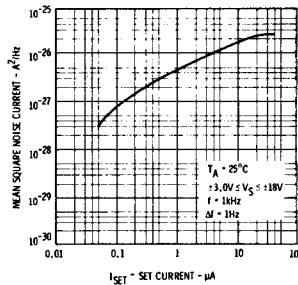
INPUT NOISE VOLTAGE
AS A FUNCTION OF
SET CURRENT



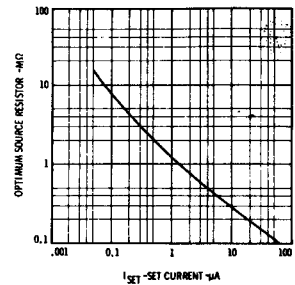
INPUT NOISE VOLTAGE AND
CURRENT AS A FUNCTION
OF FREQUENCY



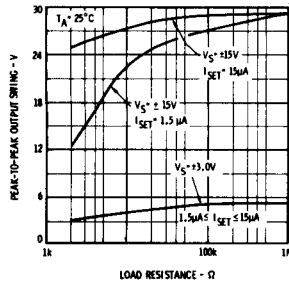
INPUT NOISE CURRENT
AS A FUNCTION OF
SET CURRENT



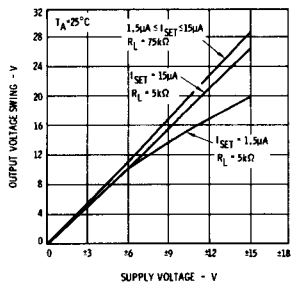
OPTIMUM SOURCE RESISTOR
FOR MINIMUM NOISE AS A
FUNCTION OF SET CURRENT



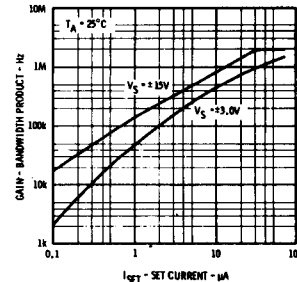
OUTPUT VOLTAGE SWING
AS A FUNCTION OF
LOAD RESISTANCE



OUTPUT VOLTAGE SWING
AS A FUNCTION OF
SUPPLY VOLTAGE

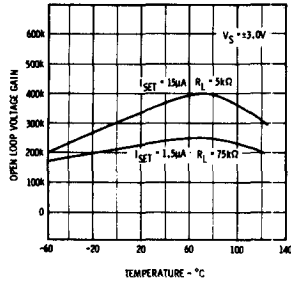


GAIN-BANDWIDTH
PRODUCT AS A FUNCTION
OF SET CURRENT

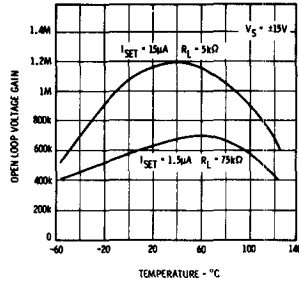


TYPICAL PERFORMANCE CURVES FOR $\mu A776$ AND $\mu A776C$

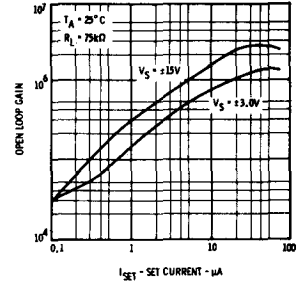
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF AMBIENT TEMPERATURE



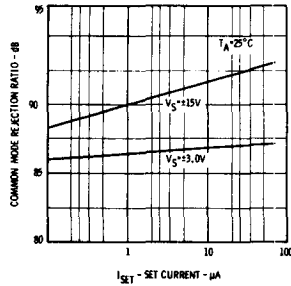
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF AMBIENT TEMPERATURE



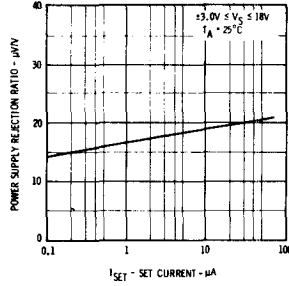
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF SET CURRENT



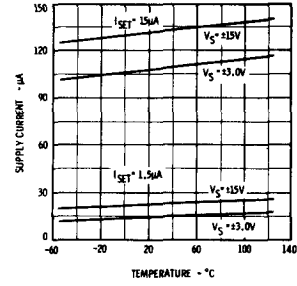
COMMON MODE REJECTION RATIO AS A FUNCTION OF SET CURRENT



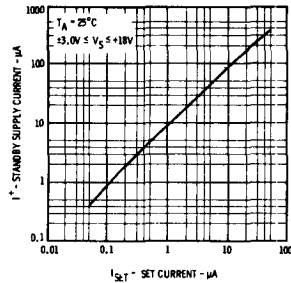
POWER SUPPLY REJECTION RATIO AS A FUNCTION OF SET CURRENT



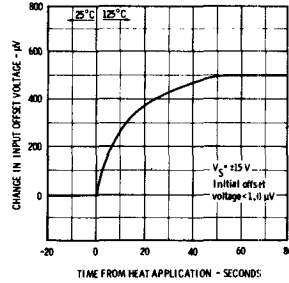
SUPPLY CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



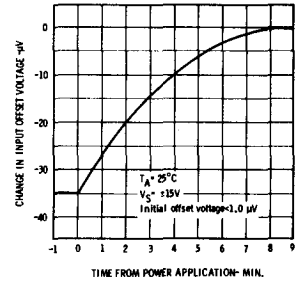
STANDBY SUPPLY CURRENT AS A FUNCTION OF SET CURRENT



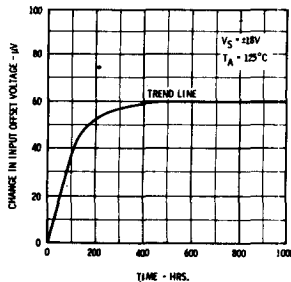
THERMAL RESPONSE OF INPUT OFFSET VOLTAGE TO STEP CHANGE OF CASE TEMPERATURE



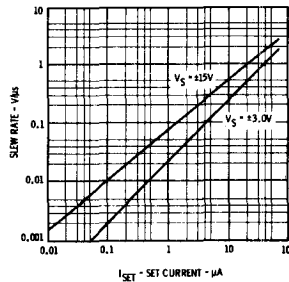
STABILIZATION TIME OF INPUT OFFSET VOLTAGE FROM POWER ON



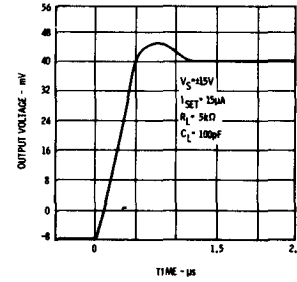
INPUT OFFSET VOLTAGE DRIFT AS A FUNCTION OF TIME



SLEW RATE AS A FUNCTION OF SET CURRENT

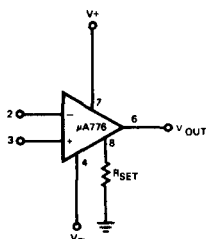


VOLTAGE FOLLOWER TRANSIENT RESPONSE (UNITY GAIN)

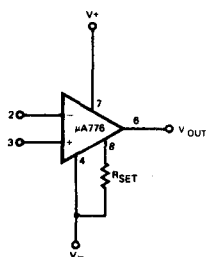


BIASING CIRCUITS

RESISTOR BIASING



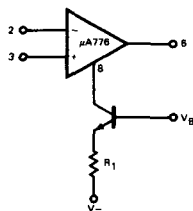
R_{SET} CONNECTED TO GROUND



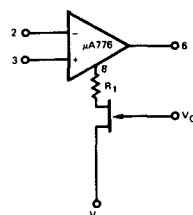
R_{SET} CONNECTED TO V^-

* Recommended for supply voltages less than $\pm 6V$.

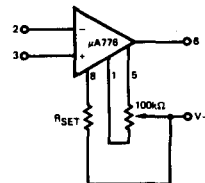
TRANSISTOR CURRENT SOURCE BIASING



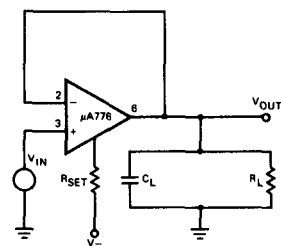
FET CURRENT SOURCE BIASING



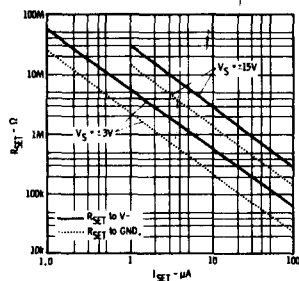
VOLTAGE OFFSET NULL CIRCUIT



TRANSIENT RESPONSE TEST CIRCUIT



SET CURRENT AS A FUNCTION OF SET RESISTOR



QUIESCENT CURRENT SETTING RESISTOR (I_{SET} TO V^-)

V_S	I_{SET}	
	$1.5\mu A$	$15\mu A$
$\pm 1.5 V$	$1.7M\Omega$	$170k\Omega$
$\pm 3.0 V$	$3.6M\Omega$	$360k\Omega$
$\pm 6.0 V$	$7.5M\Omega$	$750k\Omega$
$\pm 15 V$	$20M\Omega$	$2.0M\Omega$

Note: The $\mu A776$ may be operated with R_{SET} connected to ground or V^- .

I_{SET} EQUATIONS:

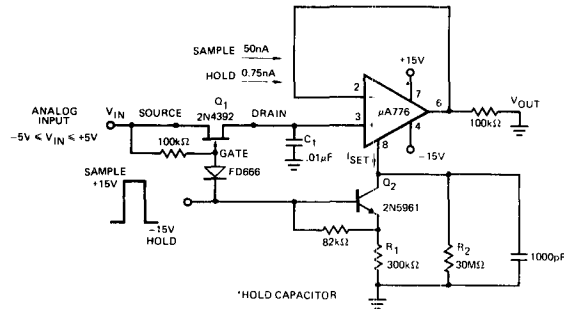
$$I_{SET} = \frac{V^+ - 0.7 - V^-}{R_{SET}}$$

where R_{SET} is connected to V^-

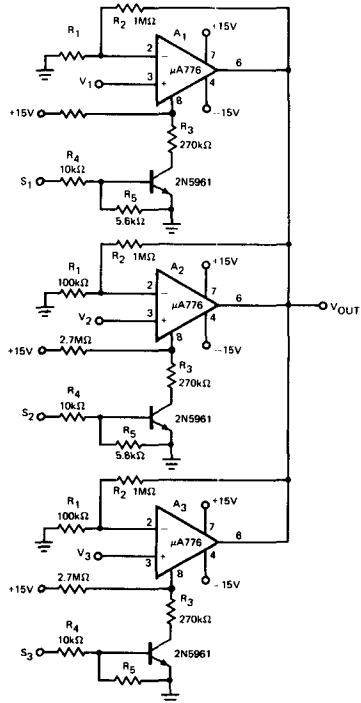
$$I_{SET} = \frac{V^+ - 0.7}{R_{SET}}$$

where R_{SET} is connected to ground.

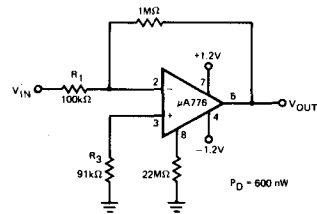
TYPICAL APPLICATIONS
HIGH ACCURACY SAMPLE AND HOLD



MULTIPLEXING AND SIGNAL CONDITIONING
WITHOUT FETs



NANO-WATT AMPLIFIER



HIGH INPUT IMPEDANCE
AMPLIFIER

